



INTERNATIONAL APPLICATION PUBLISHED UNDER THE PATENT COOPERATION TREATY (PCT)

(51) International Patent Classification ⁶ : H04B 1/707, 1/62	A1	(11) International Publication Number: WO 99/53625 (43) International Publication Date: 21 October 1999 (21.10.99)
--	-----------	--

(21) International Application Number: PCT/SE99/00490

(22) International Filing Date: 26 March 1999 (26.03.99)

(30) Priority Data:
09/056,651 8 April 1998 (08.04.98) US

(71) Applicant: TELEFONAKTIEBOLAGET LM ERICSSON
(publ) [SE/SE]; S-126 25 Stockholm (SE).

(72) Inventors: HEDBERG, Bo; Kastrupgatan 10, S-164 42 Kista (SE). HERMANSSON, Bo; Rottnerosbacken 26, S-123 48 Farsta (SE). FRANK, Georg; Kieler Strasse 26A, D-90425 Nürnberg (DE). NYSTRÖM, Christian; Flintåsvägen 12 NB, S-192 59 Sollentuna (SE).

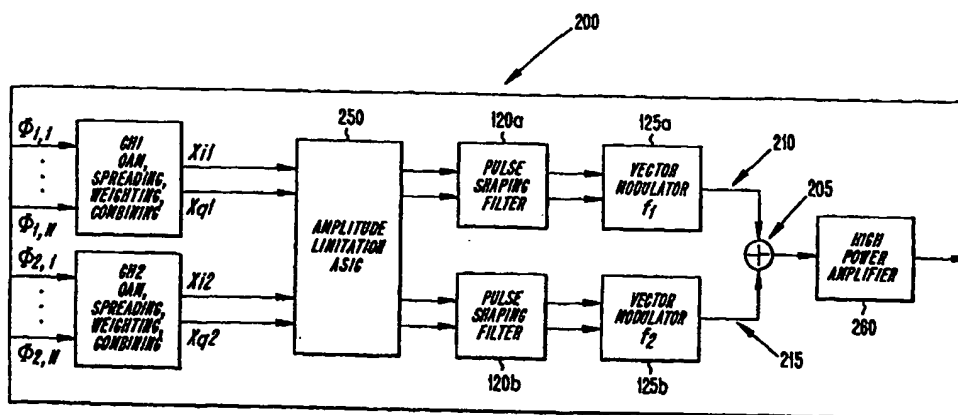
(74) Agent: ERICSSON RADIO SYSTEMS AB; Common Patent Dept., S-164 80 Stockholm (SE).

(81) Designated States: AE, AL, AM, AT, AU, AZ, BA, BB, BG, BR, BY, CA, CH, CN, CU, CZ, DE, DK, EE, ES, FI, GB, GD, GE, GH, GM, HR, HU, ID, IL, IN, IS, JP, KE, KG, KP, KR, KZ, LC, LK, LR, LS, LT, LU, LV, MD, MG, MK, MN, MW, MX, NO, NZ, PL, PT, RO, RU, SD, SE, SG, SI, SK, SL, TJ, TM, TR, TT, UA, UG, UZ, VN, YU, ZA, ZW, ARIPO patent (GH, GM, KE, LS, MW, SD, SL, SZ, UG, ZW), Eurasian patent (AM, AZ, BY, KG, KZ, MD, RU, TJ, TM), European patent (AT, BE, CH, CY, DE, DK, ES, FI, FR, GB, GR, IE, IT, LU, MC, NL, PT, SE), OAPI patent (BF, BJ, CF, CG, CI, CM, GA, GN, GW, ML, MR, NE, SN, TD, TG).

Published

With international search report.
Before the expiration of the time limit for amending the claims and to be republished in the event of the receipt of amendments.

(54) Title: AMPLITUDE LIMITATION IN A CDMA SYSTEM



(57) Abstract

In a telecommunications network that employs a CDMA scheme, the amplitude associated with each independent CDMA carrier is digitally limited, thereby limiting the peak-to-average power ratio. This, in turn, is accomplished by measuring the instantaneous amplitude for the in-phase and quadrature signals that make up each CDMA carrier, deriving a maximum amplitude based on the instantaneous amplitude measurements, and then deriving one or more scaling factors based, in-part, on maximum amplitude. The one or more scaling factors are then applied to the in-phase and quadrature signals, which are subsequently filtered, combined and modulated by a corresponding CDMA carrier frequency.

FOR THE PURPOSES OF INFORMATION ONLY

Codes used to identify States party to the PCT on the front pages of pamphlets publishing international applications under the PCT.

AL	Albania	ES	Spain	LS	Lesotho	SI	Slovenia
AM	Armenia	FI	Finland	LT	Lithuania	SK	Slovakia
AT	Austria	FR	France	LU	Luxembourg	SN	Senegal
AU	Australia	GA	Gabon	LV	Latvia	SZ	Swaziland
AZ	Azerbaijan	GB	United Kingdom	MC	Monaco	TD	Chad
BA	Bosnia and Herzegovina	GE	Georgia	MD	Republic of Moldova	TG	Togo
BB	Barbados	GH	Ghana	MG	Madagascar	TJ	Tajikistan
BE	Belgium	GN	Guinea	MK	The former Yugoslav Republic of Macedonia	TM	Turkmenistan
BF	Burkina Faso	GR	Greece	ML	Mali	TR	Turkey
BG	Bulgaria	HU	Hungary	MN	Mongolia	TT	Trinidad and Tobago
BJ	Benin	IE	Ireland	MR	Mauritania	UA	Ukraine
BR	Brazil	IL	Israel	MW	Malawi	UG	Uganda
BY	Belarus	IS	Iceland	MX	Mexico	US	United States of America
CA	Canada	IT	Italy	NE	Niger	UZ	Uzbekistan
CF	Central African Republic	JP	Japan	NL	Netherlands	VN	Viet Nam
CG	Congo	KE	Kenya	NO	Norway	YU	Yugoslavia
CH	Switzerland	KG	Kyrgyzstan	NZ	New Zealand	ZW	Zimbabwe
CI	Côte d'Ivoire	KP	Democratic People's Republic of Korea	PL	Poland		
CM	Cameroon	KR	Republic of Korea	PT	Portugal		
CN	China	KZ	Kazakstan	RO	Romania		
CU	Cuba	LC	Saint Lucia	RU	Russian Federation		
CZ	Czech Republic	LJ	Liechtenstein	SD	Sudan		
DE	Germany	LK	Sri Lanka	SE	Sweden		
DK	Denmark	LR	Liberia	SG	Singapore		
EE	Estonia						

AMPLITUDE LIMITATION IN A CDMA SYSTEM

BACKGROUND

5 The present invention relates to cellular radio telecommunications systems, and more particularly to cellular radio telecommunications systems that employ a code division multiple access (CDMA) scheme.

Cellular radio telecommunications systems employ one or more channel access schemes. One well-known channel access scheme is the code division multiple
10 access (CDMA) scheme. CDMA is well-known in the art. Unlike other channel access schemes (e.g., time division or frequency division multiple access), a number of different traffic channel signals are simultaneously transmitted in such a way that they overlap in both the time domain and the frequency domain.

In order to distinguish each traffic channel signal from the other traffic
15 channel signals, each traffic channel signal is encoded with one or more unique spreading codes, as is well-known in the art. By modulating each of the traffic channel signals with a spreading code, the sampling rate (i.e., the "chip rate") may be substantially increased in accordance with a spreading factor. For example, each traffic channel signal is modulated in accordance with a digital modulation scheme, e.g., a
20 quadrature amplitude modulation (QAM) or a phase shift keying (PSK) technique. Consequently, an in-phase and quadrature component signal is produced for each traffic channel signal. QAM and PSK are well known in the art. The in-phase and quadrature component signals associated with each of the traffic channels are then encoded using a unique spreading code sequence. The resulting in-phase and quadrature component
25 signal pairs are sampled (i.e., at the chip rate) and individually weighted. The in-phase and quadrature component signals are eventually combined to form a composite in-phase signal and a composite quadrature signal. The composite in-phase signal and the composite quadrature signal are then separately filtered by a low-pass, pulse shaping filter. Subsequent to filtering, the composite in-phase signal and the composite
30 quadrature signal are modulated by a cosine-carrier and a sine-carrier respectively and combined into a single, multicode CDMA signal. The single, multicode CDMA signal is then upconverted by a carrier frequency and the signal power associated with the

- 2 -

CDMA signal is boosted by a high power amplifier prior to transmission. At a receiving unit, the baseband signal associated with each of the traffic channel signals is extracted from the CDMA signal by demodulating and decoding the CDMA signal using the carrier frequency and the various spreading codes. Furthermore, it will be understood that in a typical cellular telecommunications system, the transmission source may, for example, be a high power base station, and the receiving entity may, for example, be a mobile station (i.e., a mobile telephone).

When there is an especially large number of traffic channel signals, it is sometimes preferable to generate two or more CDMA signals, wherein each of the two or more CDMA signals is upconverted by its own unique CDMA carrier frequency. The two or more upconverted CDMA signals are then independently amplified by a corresponding high power amplifier prior to transmission, or alternatively, the two or more upconverted CDMA signals are combined into a single, CDMA signal, which is then amplified by a single, high power amplifier prior to transmission.

As one skilled in the art will readily appreciate, CDMA substantially increases system bandwidth, which in turn, increases the network's traffic handling capacity as a whole. In addition, combining independent CDMA signals into a single CDMA signal, as described above, is advantageous in that a single high power amplifier is required rather than a separate high power amplifier for each independent CDMA signal. This is advantageous because high power amplifiers are expensive, and employing one high power amplifier in place of many will result in a substantial cost savings.

Despite the advantages associated with CDMA, combining multiple traffic channel signals and/or independent CDMA signals, in general, significantly increases the peak-to-average power ratio associated with the resulting CDMA signal. More specifically, the peak-to-average power ratio for a CDMA signal can be determined in accordance with the following relationship:

$$PR_{PTA} = PR_F + 10 * \log (N)$$

wherein PR_{PTA} represents the peak-to-average power ratio of the corresponding composite signal, PR_F represents the power ratio of the low-pass, pulse shaping filter and N represents the number of traffic channels which make up the CDMA signal.

The problem associated with large peak-to-average power ratios is that it
5 diminishes the efficiency of the high power amplifier in the transmitter. Efficiency, as one skilled in the art will readily understand, is measured in terms of the amount of output power (i.e., P_{mean}) divided by the amount of input power (i.e., $P_{dc} + P_{peak}$). As P_{peak} (i.e., peak power) increases relative to P_{mean} , the efficiency of the high power amplifier decreases.

10 One possible solution is to simply limit or clip the amplitude (i.e., P_{peak}) of the CDMA signal. Unfortunately, this is likely to result in the generation of intermodulation products and/or spectral distortions. Intermodulation products and/or spectral distortions are, in turn, likely to cause interference between the various traffic channel signals. Accordingly, this is not a preferred solution.

15 Another possible solution is to design a more complex high power amplifier, one that can tolerate and more efficiently amplify CDMA signals that exhibit large peak-to-average ratios. However, this too is not a preferred solution as the cost of high power amplifiers are generally proportional to complexity. Accordingly, this solution would result in driving up the cost of the telecommunications device that
20 houses the high power amplifier.

U.S. Patent 5,621,762 ("Miller et al.") offers yet another possible solution for the peak-to-average power ratio problem, that is to limit the peak-to-average power ratio before the soon-to-be transmitted telecommunications signal is filtered and subsequently amplified. More specifically, Miller describes a peak power
25 suppression device for reducing the peak-to-average power ratio of a single code sequence at the input of the high power amplifier. The peak power suppression device employs a digital signal processor (DSP) which receives the single code sequence, maps the code sequence onto a symbol constellation diagram, predicts an expected response from the pulse shaping filter and limits the amplitudes appearing on the
30 symbol constellation diagram in accordance with the expected response of the pulse shaping filter.

- 4 -

The primary problem with the solution offered in Miller is that peak power suppression device is designed for a non-CDMA application. Therefore, the peak power suppression device described therein is incapable of coping with the specific characteristics associated with CDMA, such as, high data bit rates, multiple traffic channel signals and/or multi-code sequences, and multiple CDMA carrier signals. For example, the peak power suppression device described in Miller is inherently slow, as evidenced by the fact that it employs a DSP, and by the fact that the DSP has the time necessary to execute a pulse shaping filter prediction algorithm. Therefore, a need exists for a telecommunications signal amplitude limitation device that is capable of limiting the peak-to-average power ratio of a telecommunications signal before it is filtered and subsequently amplified, and additionally, is capable of handling significantly higher data bit rates, multiple code sequences, and multiple CDMA carrier signals.

SUMMARY

In view of the problems identified above, it is an object of the present invention to provide the ability to effectively reduce the peak-to-average power ratio for a CDMA signal in such a way that the efficiency of the high power amplifier in the transmitter is not degraded.

It is another object of the present invention to reduce the peak-to-average power for a CDMA signal without generating intermodulation products and/or spectral distortions.

It is yet another object of the present invention to limit the peak-to-average power ratio when there are two or more independent CDMA carrier signals.

In accordance with one aspect of the invention, the foregoing and other objects are achieved by a method and/or apparatus that limits the amplitude of a complex code division multiple access (CDMA) signal. the method and/or apparatus comprises means for measuring an instantaneous amplitude for each of a plurality of digitally encoded sequences and means for generating a maximum amplitude as a function of the instantaneous amplitude measurements. The method and/or apparatus also includes means for deriving an amplitude scaling factor as a function of the

maximum amplitude and means for applying the amplitude scaling factor to each of the plurality of digitally encoded sequences. A CDMA signal is then generated based upon each of the amplitude limited, digitally encoded sequences.

In accordance with another aspect of the invention, the foregoing and
5 other objects are achieved by a method and/or apparatus for limiting the peak-to-average power ratio of a complex code division multiple access (CDMA) signal. The method and/or apparatus according to this alternative aspect of the invention comprises means for measuring the instantaneous amplitude for a first and a second composite in-phase signal and a first and a second composite quadrature signal, wherein the first and
10 the second composite in-phase signal and the first and the second composite quadrature signal are a function of a first and a second set of digitally encoded traffic channel signals. The method and/or apparatus also includes means for generating an amplitude scaling factor for the first and the second composite in-phase signal and the first and the second composite quadrature signal as a function of the measured instantaneous
15 amplitudes associated with the first and the second composite in-phase and quadrature signals. Once the amplitude scaling factor for the first and the second composite in-phase signal and the first and the second quadrature signal, the method and/or apparatus employs means for applying the amplitude scaling factor for the first and the second composite in-phase signal and the first and the second composite quadrature signal to
20 the first and the second composite in-phase signal and the first and the second composite quadrature signal respectively. A CDMA signal is then generated based on the first and the second in-phase and quadrature signals.

BRIEF DESCRIPTION OF THE DRAWINGS

25 The objects and advantages of the invention will be understood by reading the following detailed description in conjunction with the drawings in which:

FIG. 1 shows a technique for generating and amplifying a CDMA signal in accordance with the prior art;

FIG. 2 shows a technique for generating and amplifying a CDMA signal
30 in accordance with a preferred embodiment of the present invention;

FIG. 3 is a symbol constellation diagram;

FIG. 4 illustrates the amplitude limitation ASIC in accordance with a preferred embodiment of the present invention;

FIG. 5 illustrates the amplitude limitation ASIC in accordance with an alternative embodiment of the present invention; and

5 FIGS. 6A and 6B are symbol constellation diagrams.

DETAILED DESCRIPTION

The various features of the invention will now be described with respect to the figures, in which like parts are identified with the same reference characters.

10 FIG. 1 is a schematic diagram that depicts a prior technique for generating a CDMA signal 105. As illustrated, the CDMA signal 105 is generated by combining, two (or more) independent CDMA signals 110 and 115. In accordance with this prior technique, each traffic channel signal from a first set of digital traffic channel signals $\Phi_{1,1} \dots \Phi_{1,N}$ and each traffic channel signal from a second set of digital
15 traffic channel signals $\Phi_{2,1} \dots \Phi_{2,N}$ is modulated using a quadrature amplitude modulation (QAM) technique. This results in the generation of an in-phase and quadrature signal pair for each of the traffic channel signals. Each of the in-phase signals associated with the first set of traffic channel signals is then encoded using a unique spreading code, individually weighted and combined with other in-phase
20 signals, thereby generating a first composite in-phase signal X_{i1} , and each of the quadrature signals associated with the first set of traffic channel signals is likewise encoded, weighted and combined, thereby generating a first composite quadrature signal X_{q1} . Similarly, each of the in-phase signals associated with the second set of traffic channel signals is encoded, weighted and combined, thereby generating a second
25 composite in-phase signal X_{i2} , and each of the quadrature signals associated with the second set of traffic channel signals is encoded, weighted and combined, thereby generating a second composite quadrature signal X_{q2} . As illustrated in FIG. 1, the composite in-phase signal X_{i1} and the composite quadrature signals X_{q1} are then forwarded to a first pulse shaping filter 120a. Similarly, the composite in-phase signal
30 X_{i2} and the composite quadrature signals X_{q2} are forwarded to a second pulse shaping filter 120b. Next, the filtered signals are forwarded to a first and a second vector

- 7 -

modulator 125a and 125b. The vector modulator 125a modulates the composite in-phase signal X_{i1} by a cosine-carrier with a frequency f_1 and it modulates the composite quadrature signal X_{q1} by a sine-carrier also having a frequency f_1 . The vector modulator 125a then combines the modulated, composite in-phase signal X_{i1} with the modulated, composite quadrature signal X_{q1} , thereby generating the first independent CDMA signal 110. Simultaneously, the vector modulator 125b modulates the composite in-phase signal X_{i2} by a cosine-carrier with a frequency f_2 and it modulates the composite quadrature signal X_{q2} by a sine-carrier also having a frequency f_2 . The vector modulator 125b then combines the modulated, composite in-phase signal X_{i2} with the modulated, composite quadrature signal X_{q2} , thereby generating the second independent CDMA signal 115. The two independent CDMA signals 110 and 115 are then combined to form the CDMA signal 105, which is then forwarded to a high power amplifier 130 prior to transmission.

As explained above, the peak-to-average power ratio associated with the CDMA signal 105 increases as the number of traffic channel signals Φ increases, and an increase in the peak-to-average power ratio, in turn, reduces the efficiency of the high power amplifier 130. In addition, if an attempt is made to limit or clip the amplitude of the CDMA signal 105 in the high power amplifier 130 or in the transmitter (not shown) which houses the high power amplifier 130, a considerable amount of intermodulation and/or spectral distortion is likely to result.

FIG. 2 is a schematic diagram that depicts a technique 200 for generating a composite CDMA signal 205 in accordance with a preferred embodiment of the present invention. This technique is similar to the technique depicted in FIG. 1, in that the preferred embodiment also involves encoding and combining each of a first and a second plurality of digital traffic channel signals $\Phi_{1,1} \dots \Phi_{1,N}$ and $\Phi_{2,1} \dots \Phi_{2,N}$ into a first composite in-phase signal X_{i1} , a first composite quadrature signal X_{q1} , a second composite in-phase signal X_{i2} and a second composite quadrature signal X_{q2} . However, unlike the prior technique depicted in FIG. 1, the composite in-phase and quadrature signals X_{i1} , X_{q1} , X_{i2} and X_{q2} are forwarded to an amplitude limitation, application specific integrated circuit (ASIC) 250.

The ASIC 250 is a high speed hardware device that is capable of limiting the amplitude of the composite in-phase and quadrature signals X_{i1} , X_{q1} , X_{i2} and X_{q2} before the signals are forwarded to the pulse shaping filters 120a and 120b. The ASIC 250 will be described in greater detail below. The now filtered and amplitude adjusted in-phase and quadrature signals X_{i1} and X_{q1} are then modulated by a CDMA carrier with frequency f_1 and combined to form the first independent CDMA signal 210. Similarly, the now filtered and amplitude adjusted in-phase and quadrature signals X_{i2} and X_{q2} are modulated by a CDMA carrier with frequency f_2 and combined to form the second independent CDMA signal 215. The two independent CDMA carrier signals 210 and 215 are then upconverted and combined to form the CDMA signal 205. The signal power of the CDMA signal 205 is then boosted by the high power amplifier 260 prior to transmission.

In accordance with the preferred embodiment of the present invention, limiting the amplitude of a CDMA signal, for example CDMA signal 205, first requires the determination of a maximum amplitude a_1 , associated with the first independent CDMA signal 210, and a maximum amplitude a_2 , associated with the second independent CDMA signal 215. These determinations are better understood with reference to the symbol constellation diagram illustrated in FIG. 3, wherein S_1 represents the amplitude and phase corresponding with the first CDMA signal 210 and S_2 represents the amplitude and phase corresponding with the second CDMA signal 215. The maximum amplitudes a_1 and a_2 are then determined in accordance with the following relationships:

$$a_1 = | S_1 | = (X_{i1}^2 + X_{q1}^2)^{1/2} \quad (1)$$

$$a_2 = | S_2 | = (X_{i2}^2 + X_{q2}^2)^{1/2} \quad (2)$$

wherein X_{i1} , X_{q1} , X_{i2} and X_{q2} represent the instantaneous values of the composite in-phase and quadrature signals described above. However, one skilled in the art will understand that a_1 and a_2 could be approximated using equations other than equations (1) and (2) above.

- 9 -

Once the maximum amplitudes a_1 and a_2 have been determined, a_1 and a_2 are used to calculate a scaling factor "s". In accordance with the preferred embodiment, the scaling factor "s" is determined by the following relationships:

$$s = a_{\text{clip}} / a \quad (\text{if } a > a_{\text{clip}}) \quad (3)$$

$$s = 1 \quad (\text{if } a \leq a_{\text{clip}}) \quad (4)$$

wherein a_{clip} is defined as the maximum allowable amplitude value realized at the input of the pulse shaping filters 120a and 120b, and "a" represents a maximum overall amplitude. More specifically, the maximum overall amplitude "a" is given by the following relationship.

$$a = a_1 + a_2 \quad (5)$$

The scaling factor "s" is then used to limit the instantaneous amplitudes associated with the composite in-phase and the composite quadrature signals X_{i1} , X_{q1} , X_{i2} and X_{q2} .

FIG. 4 illustrates, in greater detail, the functional components associated with ASIC 250 which are needed to execute the preferred amplitude limitation technique described above. More specifically, ASIC 250 contains a maximum amplitude calculation module 405. The maximum amplitude calculation module 405 represents a high speed digital circuit that is capable of making the necessary measurements and computations to solve equations (1) and (2) above. ASIC 250 then forwards a_1 and a_2 to a scaling factor computation module 410. The scaling factor computation module 410 represents a high speed digital circuit that is capable of performing the necessary computations to solve equations (3), (4) and (5) above.

Once the scaling factor "s" is determined, the scaling factor calculation module 410 forwards the scaling factor "s" to scaling modules 415a and 415b. The scaling module 415a represents a high speed digital circuit that is capable of applying (e.g., multiplying) the scaling factor "s" to both the composite in-phase signal X_{i1} and the composite quadrature signal X_{q1} . Similarly, the scaling module 415b represents a

high speed digital circuit that is capable of applying the scaling factor "s" to both the composite in-phase signal Xi2 and the composite quadrature signal Xq2. Once the in-phase and quadrature signals Xi1, Xq1, Xi2 and Xq2 have been scaled, the ASIC 250 forwards the amplitude limited signals to the pulse shaping filters 120a and 120b, as illustrated in FIG. 2.

FIG. 5 illustrates an alternative embodiment for the ASIC 250. In accordance with this alternative embodiment, separate scaling factors s1 and s2 are computed by the scaling factor computation module 510, wherein scaling factor s1 is utilized for independently adjusting the instantaneous amplitude of the in-phase and quadrature signals Xi1 and Xq1, and the scaling factor s2 is utilized for independently adjusting the instantaneous amplitude of the in-phase and quadrature signals Xi2 and Xq2. More specifically, s1 and s2 may be determined in accordance with the following equations:

$$s1 = (a_{clip}/a1) * w_1 \quad (6)$$

$$s2 = (a_{clip}/a2) * w_2 \quad (7)$$

wherein w1 and w2 represent a first and a second weighting factor for independently adjusting the scaling factors s1 and s2 respectively.

The alternative technique illustrated in FIG. 5 may be employed when there is a significant disparity between the signal power levels associated with the traffic channel signals of CH1 in FIG. 2 as compared to the signal power levels associated with the traffic channel signals of CH2. If, for example, the signal power levels associated with the traffic channel signals of CH1 are significantly lower than those associated with the traffic channel signals of CH2, it may be appropriate to scale only the instantaneous amplitudes for the composite in-phase and quadrature signals Xi2 and Xq2. This can effectively be accomplished by setting the weighting factor w2 to the value "1", and by setting the weighting factor w1 such that s1 approximates the value "1". Of course, it will be understood that weighting factors w1 and w2 could be

- 11 -

set to any value that is deemed appropriate to scale the instantaneous amplitudes for the composite in-phase and quadrature signals X_{i1} , X_{q1} , X_{i2} and X_{q2} .

In accordance with yet another alternative embodiment, the instantaneous amplitude samples associated with the composite in-phase and quadrature signal samples (e.g., X_{i1} , X_{q1} , X_{i2} , X_{q2}) may be limited or clipped if the amplitude sample exceeds a predetermined maximum value. In order to prevent a corresponding decrease in the average power level of the composite CDMA signal, and hence, an undesirable increase in the PR_{PTA} of the composite CDMA signal, this alternative embodiment generates a scaling factor which is then used to increase the amplitude of one or more subsequent, composite in-phase and quadrature signal samples, wherein the increase in amplitude over the one or more subsequent samples is proportional to the decrease in amplitude of the sample that was previously clipped. Of course, adjusting the amplitude of these subsequent samples compensates for the instantaneous amplitude sample that was previously clipped. Moreover, one skilled in the art will appreciate that lower bit error rates can be achieved by modestly increasing the amplitude of several, subsequent, composite in-phase and quadrature signal samples rather than dramatically increasing the amplitude of a single, subsequent sample. This is especially true if increasing the amplitude of the single, subsequent sample results in that amplitude exceeding the aforementioned predetermined maximum value.

FIG. 6 illustrates two symbol constellations diagrams 605 and 610. The symbol constellation diagram 605 shows the location of the symbols (i.e., instantaneous amplitudes) associated with a CDMA signal (e.g., CDMA signal 205) when digital amplitude limitation, in accordance with the preferred embodiment of the present invention, is employed. The symbol constellation diagram 610 shows the location of the symbols associated with the CDMA signal when digital amplitude limitation is not employed. As one skilled in the art will readily appreciate, the transmitted symbols are all located within a circular region whose radius is defined by a_{clip} , when digital amplitude limitation is employed. However, the transmitted symbols are not necessarily located within this circular region when digital amplitude limitation is not employed. The latter case is likely to result in larger peak-to-average power ratios and, as explained above, poor high power amplifier efficiency.

-12-

The present invention has now been described with reference to several exemplary embodiments. However, it will be readily apparent to those skilled in the art that it is possible to embody the invention in specific forms other than those of the exemplary embodiments described above. This may be done without departing from
5 the spirit of the invention. These exemplary embodiments are merely illustrative and should not be considered restrictive in any way. The scope of the invention is given by the appended claims, rather than the preceding description, and all variations and equivalents which fall within the range of the claims are intended to be embraced therein.

WHAT IS CLAIMED IS:

1. An apparatus for limiting the amplitude of a complex code division multiple access (CDMA) signal characterized by:
 - 5 means for measuring an instantaneous amplitude for each of a plurality of digitally encoded sequences;
means for generating a maximum amplitude as a function of the instantaneous amplitude measurements;
means for deriving an amplitude scaling factor as a function of the
10 maximum amplitude;
means for applying the amplitude scaling factor to each of the plurality of digitally encoded sequences; and
means for generating a CDMA signal based upon each of the amplitude limited, digitally encoded sequences.
- 15 2. An apparatus in accordance with claim 1, wherein said means for deriving the amplitude scaling factor is also a function of a clipping amplitude.
3. An apparatus in accordance with claim 2, wherein the clipping amplitude
20 is a function of a pulse shaping filter.
4. An apparatus in accordance with claim 1, wherein said means for generating a CDMA signal based upon each of the amplitude limited, digitally encoded sequences comprises:
 - 25 means for filtering each of the amplitude limited, digitally encoded sequences; and
means for combining the amplitude limited, digitally encoded sequences.
5. An apparatus in accordance with claim 4 further comprising:
 - 30 means for modulating the amplitude limited, digitally encoded sequences.

6. An apparatus for limiting the peak-to-average power ratio of a complex code division multiple access (CDMA) signal characterized by:

means for measuring instantaneous amplitude for a first composite in-phase signal and a first composite quadrature signal, wherein the first composite in-phase signal and the first composite quadrature signal are a function of a first set of
5 digitally encoded traffic channel signals;

means for measuring instantaneous amplitude for a second composite in-phase signal and a second composite quadrature signal, wherein the second composite in-phase signal and the second composite quadrature signal are a function of a second
10 set of digitally encoded traffic channel signals;

means for generating an amplitude scaling factor for the first composite in-phase signal and the first composite quadrature signal as a function of the measured instantaneous amplitudes associated with the first composite in-phase and quadrature signals and the second composite in-phase and quadrature signals;

15 means for generating an amplitude scaling factor for the second composite in-phase signal and the second composite quadrature signal as a function of the measured instantaneous amplitudes associated with the first composite in-phase and quadrature signals and the second composite in-phase and quadrature signals;

20 means for applying the amplitude scaling factor for the first composite in-phase signal and the first composite quadrature signal to the first composite in-phase signal and the first composite quadrature signal;

means for applying the amplitude scaling factor for the second composite in-phase signal and the second composite quadrature signal to the second composite in-phase signal and the second composite quadrature signal; and

25 means for generating a CDMA signal based on the first and the second in-phase and quadrature signals.

7. The apparatus in accordance with claim 6 further comprising:

30 means for generating a maximum amplitude as a function of the measured instantaneous amplitudes associated with the first composite and the second

composite in-phase and quadrature signals, wherein said means for generating the amplitude scaling factor for the first composite in-phase signal and the first composite quadrature signal and said means for generating the amplitude scaling factor for the second composite in-phase signal and the second composite quadrature signal include
5 means for generating the amplitude scaling factors as a function of the maximum amplitude.

8. The apparatus of claim 7, wherein the amplitude scaling factor generated for the first composite in-phase signal and the first composite quadrature signal is
10 equivalent to the amplitude scaling factor generated for the second composite in-phase signal and the second composite quadrature signal.

9. The apparatus of claim 6 further comprising:
means for applying a first weighting factor to the amplitude scaling
15 factor generated for the first composite in-phase signal and the first composite quadrature signal; and
means for applying a second weighting factor to the amplitude scaling factor generated for the second composite in-phase signal and the second composite quadrature signal.

20 10. The apparatus of claim 9 further comprising:
means for adjusting the first and the second weighting factors such that the amplitude scaling factor generated for the first composite in-phase signal and the first composite quadrature signal is equivalent to said amplitude scaling factor
25 generated for the second composite in-phase signal and the second composite quadrature signal.

11. The apparatus of claim 9 further comprising:
means for adjusting the first weighting factor such that the amplitude
30 scaling factor generated for the first composite in-phase signal and the first composite

- 16 -

quadrature signal is a function of a signal power level representative of the first set of digitally encoded traffic channel signals; and

means for adjusting the second weighting factor, independent of the means for adjusting the first weighting factor, such that the amplitude scaling factor generated for the second composite in-phase signal and the second composite quadrature signal is a function of a signal power level representative of the second set of digitally encoded traffic channel signals.

12. The apparatus in accordance with claim 6 further comprising:
means for maintaining average power of the CDMA signal by applying an amplitude scaling factor to one or more composite in-phase and composite quadrature signal samples, wherein the amplitude of the one or more composite in-phase and composite quadrature signal samples is increased to compensate for a corresponding decrease in amplitude of a previous composite in-phase and composite quadrature signal sample.

13. The apparatus of claim 6 further comprising:
means for filtering the first composite in-phase signal and the first composite quadrature signal;
means for filtering the second composite in-phase signal and the second composite quadrature signal;
means for modulating the filtered, first composite in-phase signal by a first cosine-carrier and the filtered, first composite quadrature signal by a first sine-carrier, wherein the first sine-carrier has a same frequency as the first cosine-carrier;
means for modulating the filtered, second composite in-phase signal by a second cosine-carrier and the filtered, second composite quadrature signal by a second sine-carrier, wherein the second sine-carrier has a same frequency as the second cosine-carrier;
means for combining the first filtered, composite in-phase signal with the first filtered, composite quadrature signal, thereby generating a first independent CDMA signal;

- 17 -

means for combining the second filtered, composite in-phase signal with the second filtered, composite quadrature signal, thereby generating a second independent CDMA signal;

means for upconverting the first independent CDMA signal with a first
5 CDMA carrier frequency; and

means for upconverting the second independent CDMA signal with a second CDMA carrier frequency.

14. The apparatus in accordance with claim 13, wherein said means for
10 generating a CDMA signal comprises:

means for combining the first independent CDMA signal with the second independent CDMA signal.

15. A method for limiting the amplitude of a complex code division multiple
15 access (CDMA) signal comprising the steps of:

measuring an instantaneous amplitude for each of a plurality of digitally encoded sequences;

generating a maximum amplitude as a function of the instantaneous amplitude measurements;

20 deriving an amplitude scaling factor as a function of the maximum amplitude;

applying the amplitude scaling factor to each of the plurality of digitally encoded sequences; and

25 generating a CDMA signal based upon each of the amplitude limited, digitally encoded sequences.

16. A method in accordance with claim 15, wherein said step of deriving the amplitude scaling factor is also a function of a clipping amplitude.

30 17. A method in accordance with claim 16, wherein the clipping amplitude is a function of a pulse shaping filter.

18. A method in accordance with claim 15, wherein said step of generating a CDMA signal based upon each of the amplitude limited, digitally encoded sequences comprises the steps of:
- 5 filtering each of the amplitude limited, digitally encoded sequences; and
combining the amplitude limited, digitally encoded sequences.
19. A method in accordance with claim 18 further comprising the step of:
modulating the amplitude limited, digitally encoded sequences.
- 10 20. A method for limiting the peak-to-average power ratio of a complex code division multiple access (CDMA) signal comprising the steps of:
- measuring instantaneous amplitude for a first composite in-phase signal
and a first composite quadrature signal, wherein the first composite in-phase signal and
the first composite quadrature signal are a function of a first set of digitally encoded
15 traffic channel signals;
- measuring instantaneous amplitude for a second composite in-phase
signal and a second composite quadrature signal, wherein the second composite in-
phase signal and the second composite quadrature signal are a function of a second set
of digitally encoded traffic channel signals;
- 20 generating an amplitude scaling factor for the first composite in-phase
signal and the first composite quadrature signal as a function of the measured
instantaneous amplitudes associated with the first composite in-phase and quadrature
signals and the second in-phase and quadrature signals;
- generating an amplitude scaling factor for the second composite in-phase
25 signal and the second composite quadrature signal as a function of the measured
instantaneous amplitudes associated with the first composite in-phase and quadrature
signals and the second composite in-phase and quadrature signals;
- applying the amplitude scaling factor for the first composite in-phase
signal and the first composite quadrature signal to the first composite in-phase signal
30 and the first composite quadrature signal;

applying the amplitude scaling factor for the second composite in-phase signal and the second composite quadrature signal to the second composite in-phase signal and the second composite quadrature signal; and

generating a CDMA signal based on the first and the second in-phase
5 and quadrature signals.

21. A method in accordance with claim 20 further comprising the steps of:
generating a maximum amplitude as a function of the measured
instantaneous amplitudes associated with the first composite and the second composite
10 in-phase and quadrature signals, wherein the amplitude scaling factor for the first
composite in-phase and quadrature signals and amplitude scaling factor for the second
composite in-phase and quadrature signals are also generated as a function of the
maximum amplitude.

15 22. A method in accordance with claim 21, wherein the amplitude scaling
factor generated for the first composite in-phase signal and the first composite
quadrature signal is equivalent to the amplitude scaling factor generated for the second
composite in-phase signal and the second composite quadrature signal.

20 23. A method in accordance with claim 20 further comprising the steps of:
applying a first weighting factor to the amplitude scaling factor generated
for the first composite in-phase signal and the first composite quadrature signal; and
applying a second weighting factor to the amplitude scaling factor
generated for the second composite in-phase signal and the second composite
25 quadrature signal.

24. A method in accordance with claim 23 further comprising the step of:
adjusting the first and the second weighting factors such that the
amplitude scaling factor generated for the first composite in-phase signal and the first
30 composite quadrature signal is equivalent to said amplitude scaling factor generated for
the second composite in-phase signal and the second composite quadrature signal.

- 20 -

25. A method in accordance with claim 23 further comprising the steps of:
adjusting the first weighting factor such that the amplitude scaling factor
generated for the first composite in-phase signal and the first composite quadrature
signal is a function of a signal power level representative of the first set of digitally
5 encoded traffic channel signals; and
adjusting the second weighting factor, independent of the means for
adjusting the first weighting factor, such that the amplitude scaling factor generated for
the second composite in-phase signal and the second composite quadrature signal is a
function of a signal power level representative of the second set of digitally encoded
10 traffic channel signals.
26. A method in accordance with claim 20 further comprising the step of:
maintaining average power of the CDMA signal by applying the
amplitude scaling factor to one or more composite in-phase and composite quadrature
15 signal samples, wherein the amplitude of the one or more composite in-phase and
composite quadrature signal samples is increased to compensate for a corresponding
decrease in amplitude of a previous composite in-phase and composite quadrature signal
sample.
- 20 27. A method in accordance with claim 20 further comprising the steps of:
filtering the first composite in-phase signal and the first composite
quadrature signal;
filtering the second composite in-phase signal and the second composite
quadrature signal;
25 modulating the filtered, first composite in-phase signal by a first cosine-
carrier and the filtered, first composite quadrature signal by a first sine-carrier, wherein
the first sine-carrier has a same frequency as the first cosine-carrier;
modulating the filtered, second composite in-phase signal by a second
cosine-carrier and the filtered, second composite quadrature signal by a second sine-
30 carrier, wherein the second sine-carrier has a same frequency as the second cosine-
carrier;

-21-

combining the first filtered, composite in-phase signal with the first filtered, composite quadrature signal, thereby generating a first independent CDMA signal;

5 combining the second filtered, composite in-phase signal with the second filtered, composite quadrature signal, thereby generating a second independent CDMA signal;

upconverting the first independent CDMA signal with a first CDMA carrier frequency; and

10 upconverting the second independent CDMA signal with a second CDMA carrier frequency.

28. A method in accordance with claim 27, wherein said step of generating a CDMA signal comprises the step of:

15 combining the first independent CDMA signal with the second independent CDMA signal.

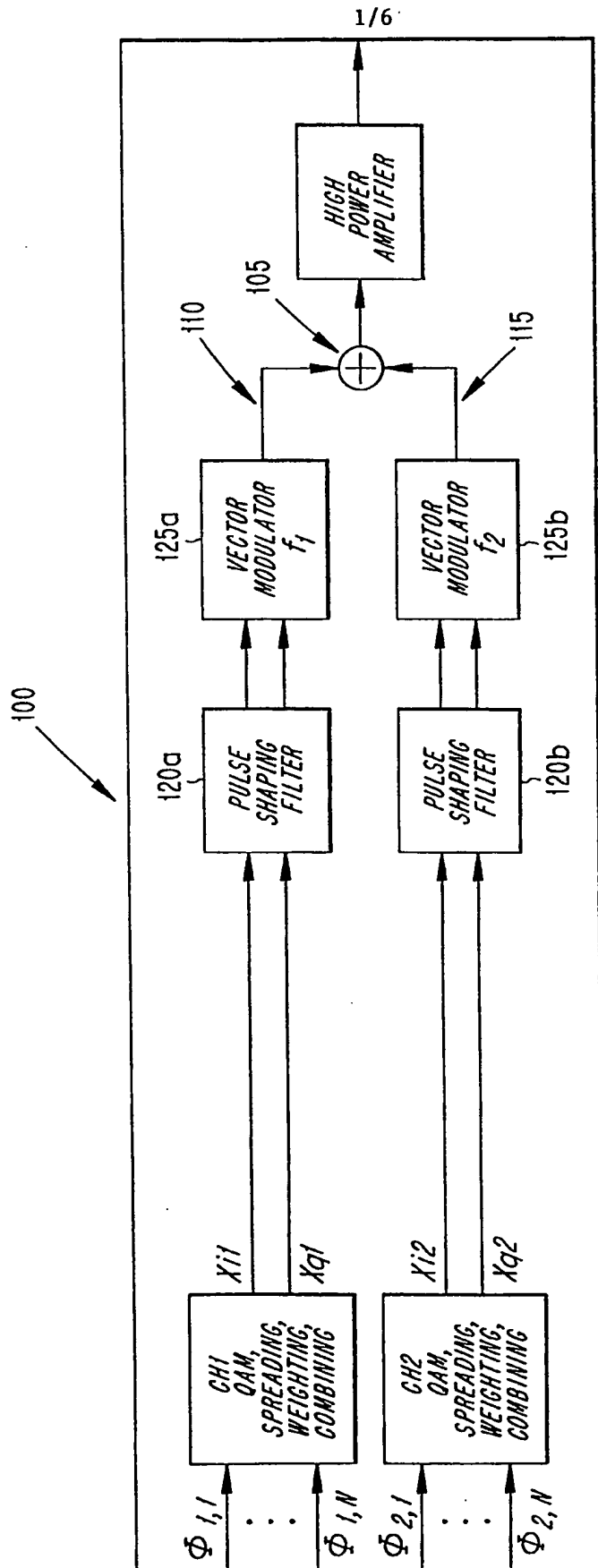


FIG. 1
(PRIOR ART)

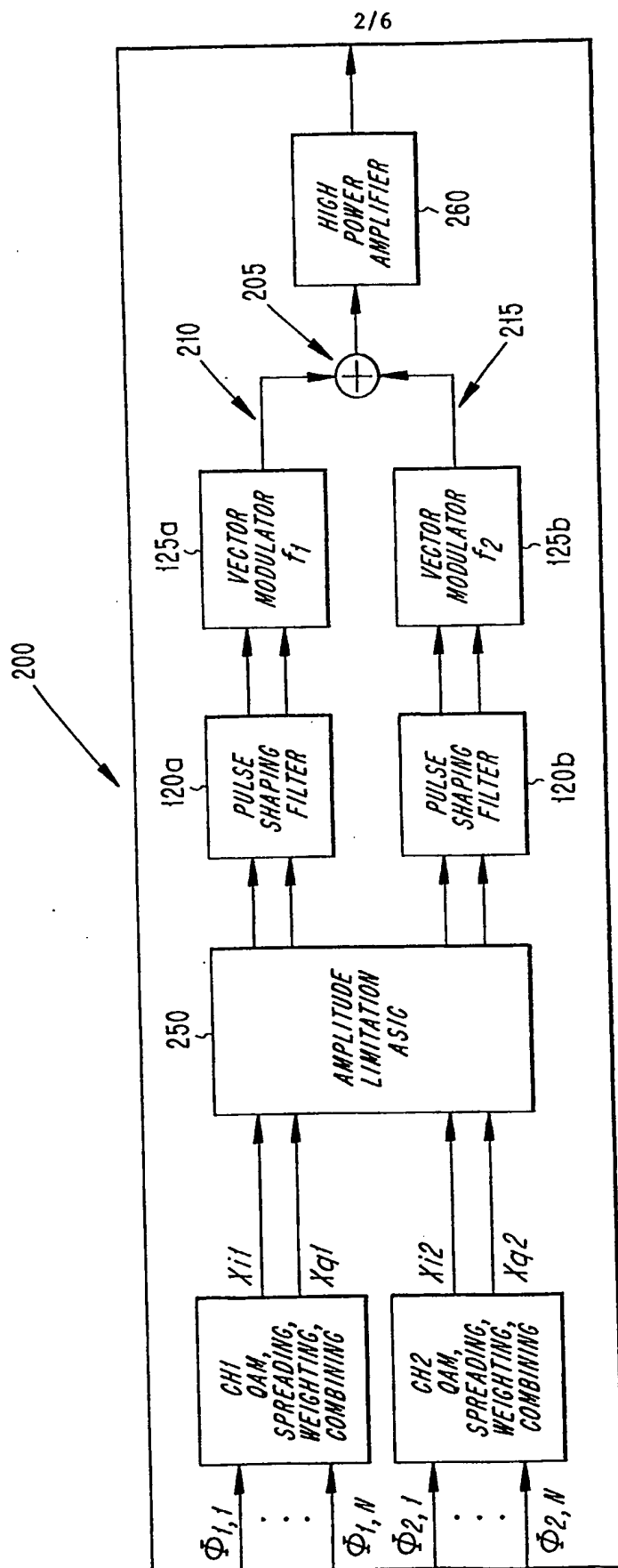


FIG. 2

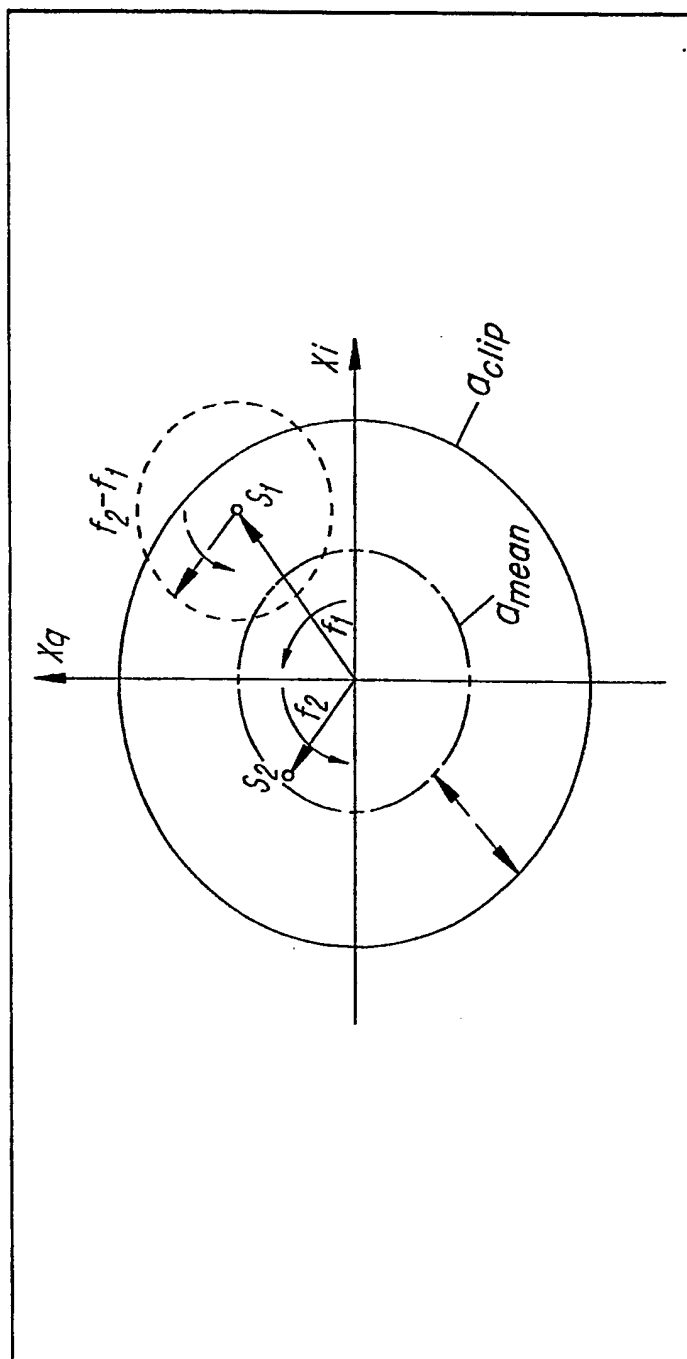


FIG. 3

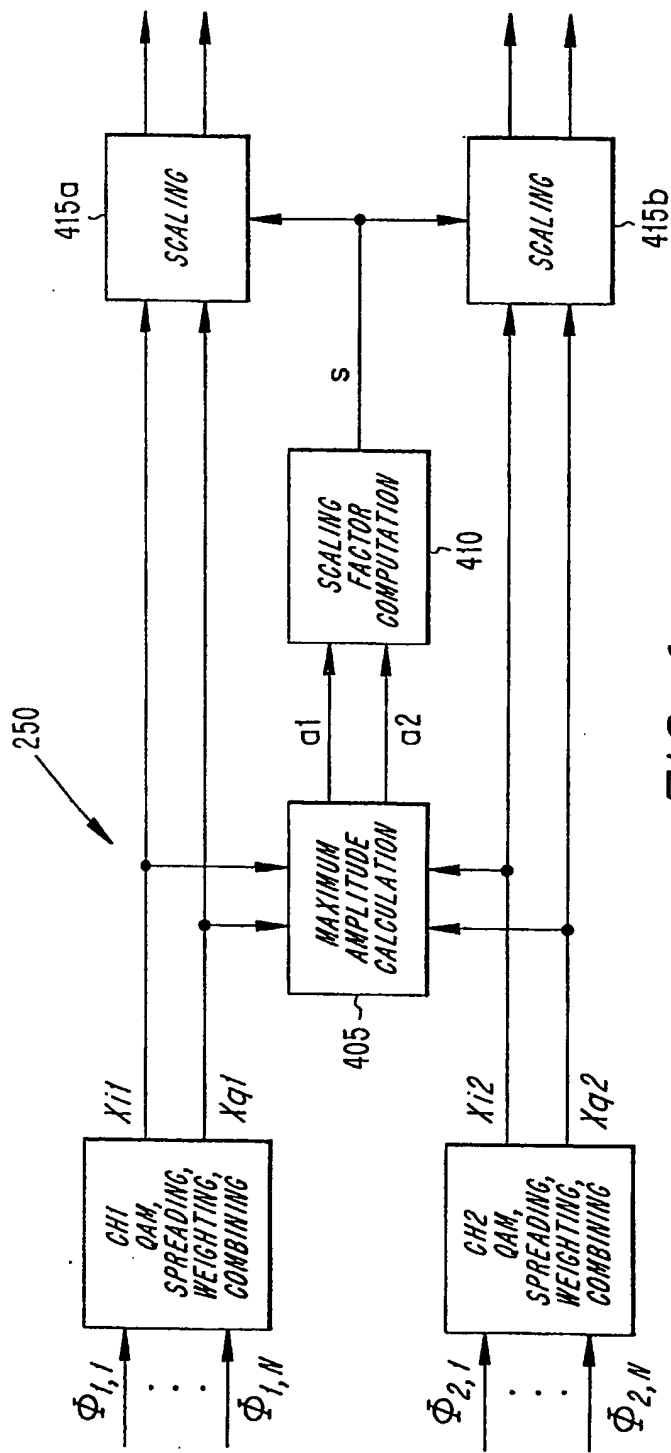


FIG. 4

250

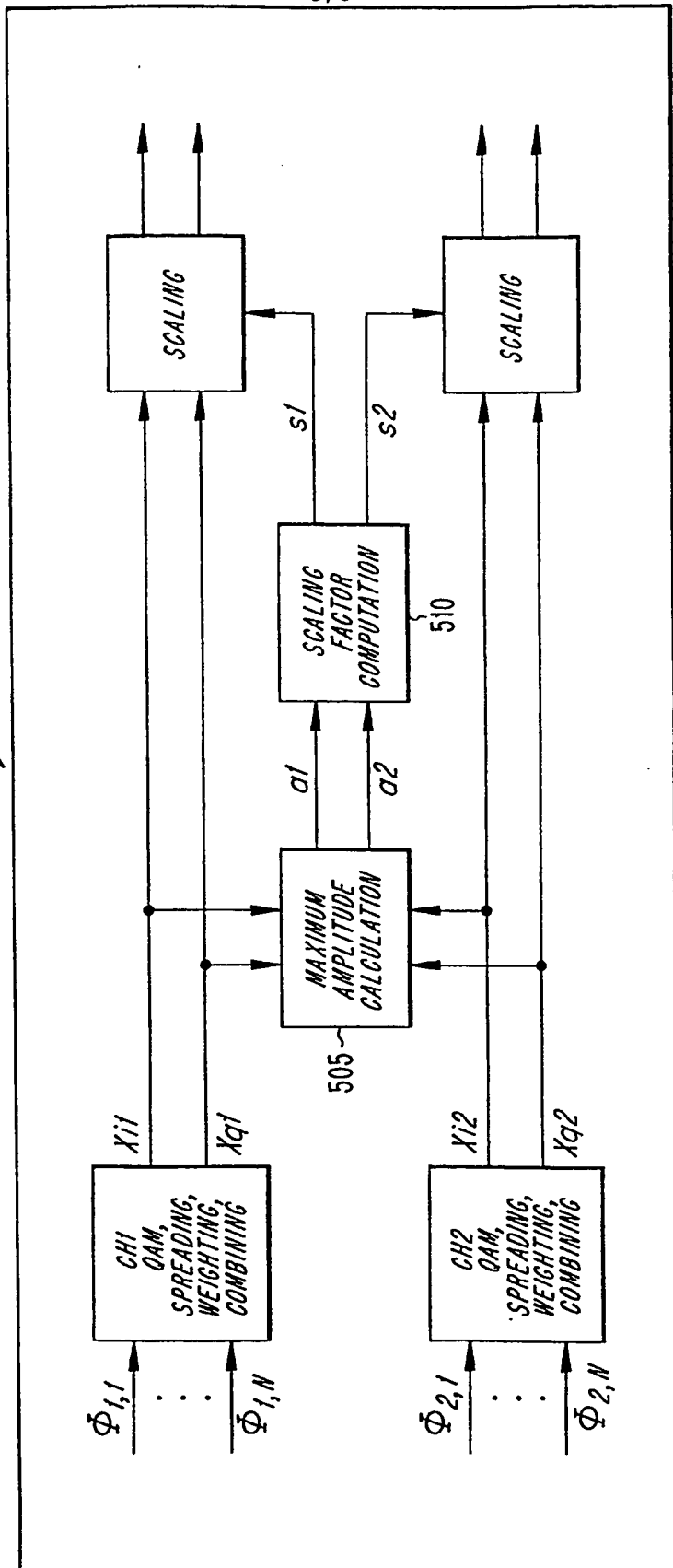


FIG. 5

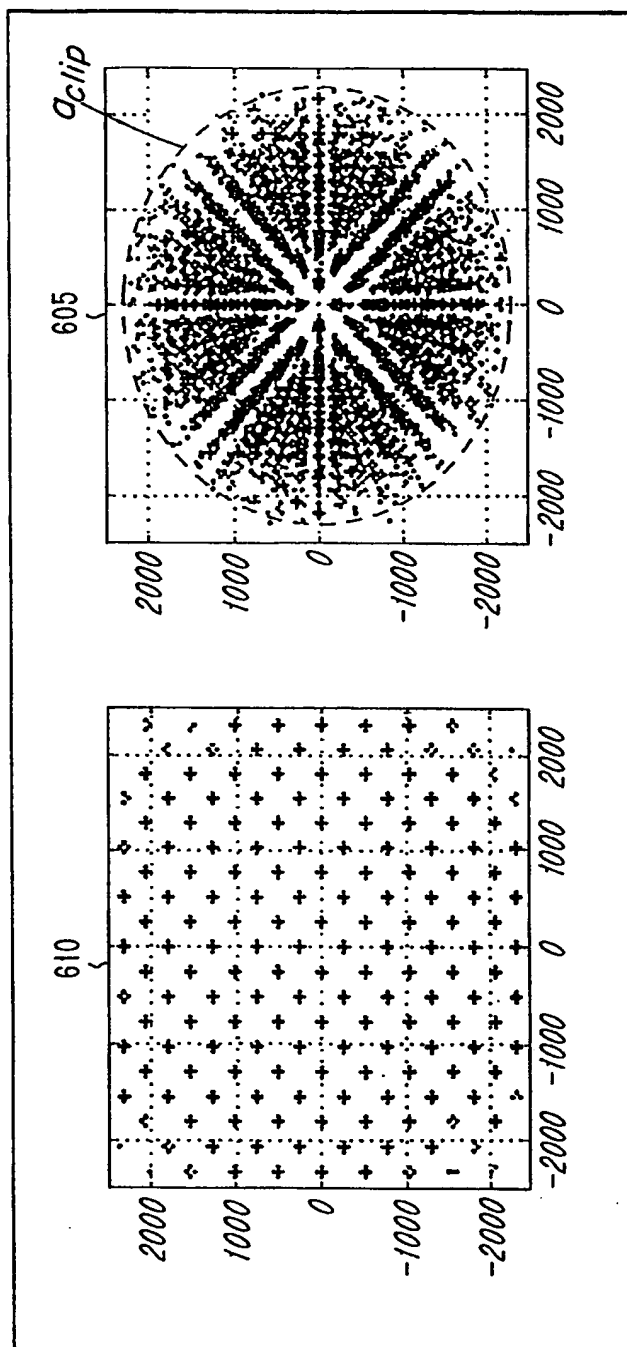


FIG. 6

INTERNATIONAL SEARCH REPORT

International Application No

PCT/SE 99/00490

A. CLASSIFICATION OF SUBJECT MATTER
IPC 6 H04B1/707 H04B1/62

According to International Patent Classification (IPC) or to both national classification and IPC

B. FIELDS SEARCHED

Minimum documentation searched (classification system followed by classification symbols)

IPC 6 H04B H04J

Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched

Electronic data base consulted during the international search (name of data base and, where practical, search terms used)

C. DOCUMENTS CONSIDERED TO BE RELEVANT

Category *	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
A	WO 96 38944 A (DSC COMMUNICATIONS) 5 December 1996 (1996-12-05) page 3, line 6 - page 5, line 5 page 34, line 8 - page 37, line 12; figure 19	1,2,4-6, 13,15, 16, 18-20,27
A	WO 96 36144 A (UNISYS CORP) 14 November 1996 (1996-11-14) page 2, line 28 - page 4, line 17 page 6, line 13 - page 8, line 33; figures 1,2	1,5,6, 13,15, 19,20,27
	-/-	

☒ Further documents are listed in the continuation of box C.

☒ Patent family members are listed in annex.

* Special categories of cited documents :

"A" document defining the general state of the art which is not considered to be of particular relevance

"E" earlier document but published on or after the international filing date

"L" document which may throw doubts on priority claim(s) or which is cited to establish the publication date of another citation or other special reason (as specified)

"O" document referring to an oral disclosure, use, exhibition or other means

"P" document published prior to the international filing date but later than the priority date claimed

"T" later document published after the international filing date or priority date and not in conflict with the application but cited to understand the principle or theory underlying the invention

"X" document of particular relevance; the claimed invention cannot be considered novel or cannot be considered to involve an inventive step when the document is taken alone

"Y" document of particular relevance; the claimed invention cannot be considered to involve an inventive step when the document is combined with one or more other such documents, such combination being obvious to a person skilled in the art.

"&" document member of the same patent family

Date of the actual completion of the international search

26 August 1999

Date of mailing of the international search report

03/09/1999

Name and mailing address of the ISA

European Patent Office, P.B. 5818 Patentlaan 2
NL - 2280 HV Rijswijk
Tel. (+31-70) 340-2040, Tx. 31 651 epo nl,
Fax: (+31-70) 340-3016

Authorized officer

Bossen, M

INTERNATIONAL SEARCH REPORT

International Application No

PCT/SE 99/00490

C.(Continuation) DOCUMENTS CONSIDERED TO BE RELEVANT

Category *	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
A	<p>US 5 668 806 A (ARAI YASUYUKI ET AL) 16 September 1997 (1997-09-16)</p> <p>column 1, line 44 - line 58 column 2, line 55 - column 3, line 19; figure 1 column 5, line 48 - line 65; figure 10A</p>	<p>1,5,6, 13,15, 19,20,27</p>
A	<p>US 5 621 762 A (MILLER SCOTT L ET AL) 15 April 1997 (1997-04-15) cited in the application column 2, line 53 - column 4, line 44; figures 3-5</p>	

INTERNATIONAL SEARCH REPORT

Information on patent family members

International Application No

PCT/SE 99/00490

Patent document cited in search report	Publication date	Patent family member(s)	Publication date
WO 9638944 A	05-12-1996	GB 2301751 A	11-12-1996
		AU 2245799 A	20-05-1999
		AU 2245899 A	27-05-1999
		AU 5972896 A	18-12-1996
		AU 5973296 A	18-12-1996
		AU 5973396 A	18-12-1996
		AU 5973796 A	18-12-1996
		AU 5973896 A	24-12-1996
		AU 5974896 A	18-12-1996
		AU 5975096 A	18-12-1996
		AU 5975896 A	18-12-1996
		AU 5977896 A	18-12-1996
		AU 5984896 A	18-12-1996
		AU 5985396 A	18-12-1996
		AU 6032896 A	18-12-1996
		AU 6033896 A	18-12-1996
		AU 6034896 A	18-12-1996
		AU 6034996 A	18-12-1996
		AU 6037696 A	18-12-1996
		AU 6037796 A	18-12-1996
		AU 6044296 A	18-12-1996
		AU 6148996 A	18-12-1996
		AU 705738 B	03-06-1999
		AU 6475896 A	18-12-1996
		BR 9608335 A	05-01-1999
		BR 9608346 A	05-01-1999
		BR 9608347 A	05-01-1999
		BR 9609468 A	02-03-1999
		CA 2222705 A	05-12-1996
		CA 2222734 A	05-12-1996
		CN 1194071 A	23-09-1998
		CN 1191045 A	19-08-1998
		CN 1192831 A	09-09-1998
		DE 69601795 D	22-04-1999
		EP 0829150 A	18-03-1998
		EP 0829174 A	18-03-1998
		EP 0834123 A	08-04-1998
		EP 0830763 A	25-03-1998
		EP 0830748 A	25-03-1998
		EP 0830749 A	25-03-1998
		EP 0830750 A	25-03-1998
		EP 0872026 A	21-10-1998
		EP 0830759 A	25-03-1998
		EP 0830775 A	25-03-1998
		EP 0830751 A	25-03-1998
		EP 0872062 A	21-10-1998
		ES 2129972 T	16-06-1999
		GB 2301736 A	11-12-1996
		GB 2301737 A	11-12-1996
		GB 2302240 A	08-01-1997
WO 9636144 A	14-11-1996	US 5793797 A	11-08-1998
		AU 5678896 A	29-11-1996
		CA 2220393 A	14-11-1996
		EP 0824810 A	25-02-1998
US 5668806 A	16-09-1997	JP 8088588 A	02-04-1996

INTERNATIONAL SEARCH REPORT

Information on patent family members

Int: lional Application No

PCT/SE 99/00490

Patent document cited in search report	Publication date	Patent family member(s)	Publication date
US 5621762 A	15-04-1997	AU 702758 B	04-03-1999
		AU 5929496 A	09-01-1997
		CA 2224572 A	27-12-1996
		CN 1191647 A	26-08-1998
		CZ 9704021 A	13-05-1998
		DE 19681454 T	20-08-1998
		EP 0883953 A	16-12-1998
		GB 2317795 A	01-04-1998
		NZ 309293 A	29-04-1999
		SE 9704633 A	05-02-1998
		WO 9642161 A	27-12-1996